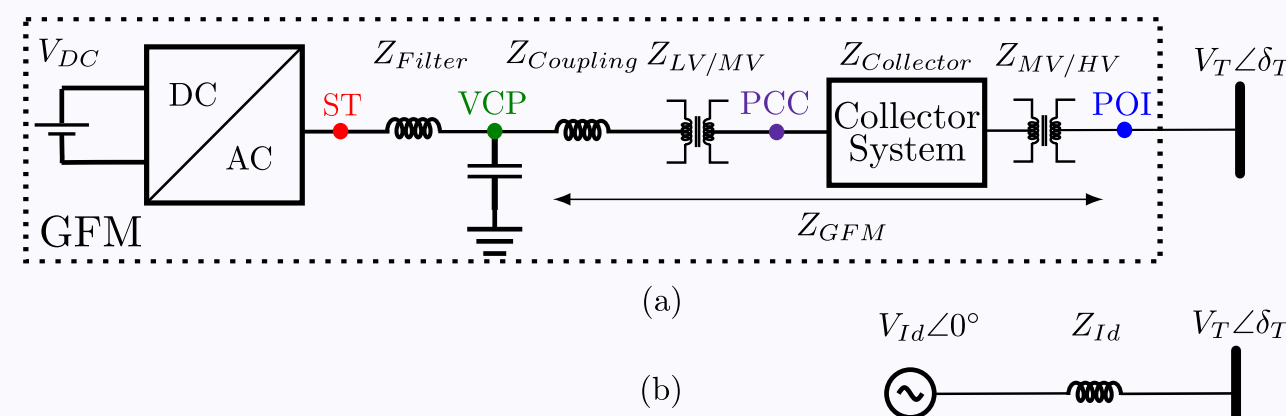


Equivalent Circuit Modeling of GFM Inverters in (Sub)-Transient Time

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I. GFM Definition & Compliance

- Definition:** A GFM should maintain a nearly constant internal voltage source (IVS) behind an effective impedance (Z_{Eff}), a.k.a. a Thevenin Source in (sub)-transient time.
- Location of IVS? → ST / VCP / None?**



- IVS location: at VCP (filter capacitor), not ST.
- IVS behind $Z_{Eff} = Z_{GFM} + Z_{Control}$.

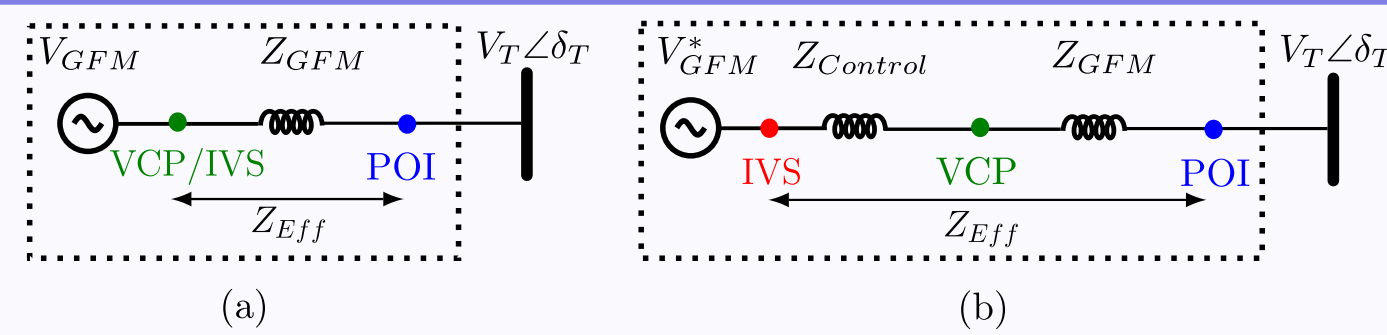


Figure 2: Equivalent GFM plant (a) without virtual impedance, (b) with virtual $Z_{Control}$.

- ENTSO-E recommends a permissible range for values of X_{Eff} ; $R_{Eff}/X_{Eff} = 0.1$.

Table 1: ENTSO-E's proposed range of values of X_{Eff} (at 50 Hz) for different definitions of the POI.

Location of POI	Min (p.u.)	Default (p.u.)	Max (p.u.)
LV terminals	0.17	0.25	0.27
MV terminals	0.25	0.33	0.35
HV terminals	0.40	0.48	0.50

- Open question:** How to quantify Z_{Eff} for a black-box GFM?

II. Admittance Spectra

- $Y_{qd}/Y_{dq} \rightarrow$ Reactive/Active power response to a voltage magnitude/phase perturbation.

$$\begin{bmatrix} I_d(s) \\ I_q(s) \end{bmatrix} = \begin{bmatrix} Y_{dd}(s) & Y_{dq}(s) \\ Y_{qd}(s) & Y_{qq}(s) \end{bmatrix} \begin{bmatrix} V_d(s) \\ V_q(s) \end{bmatrix}$$

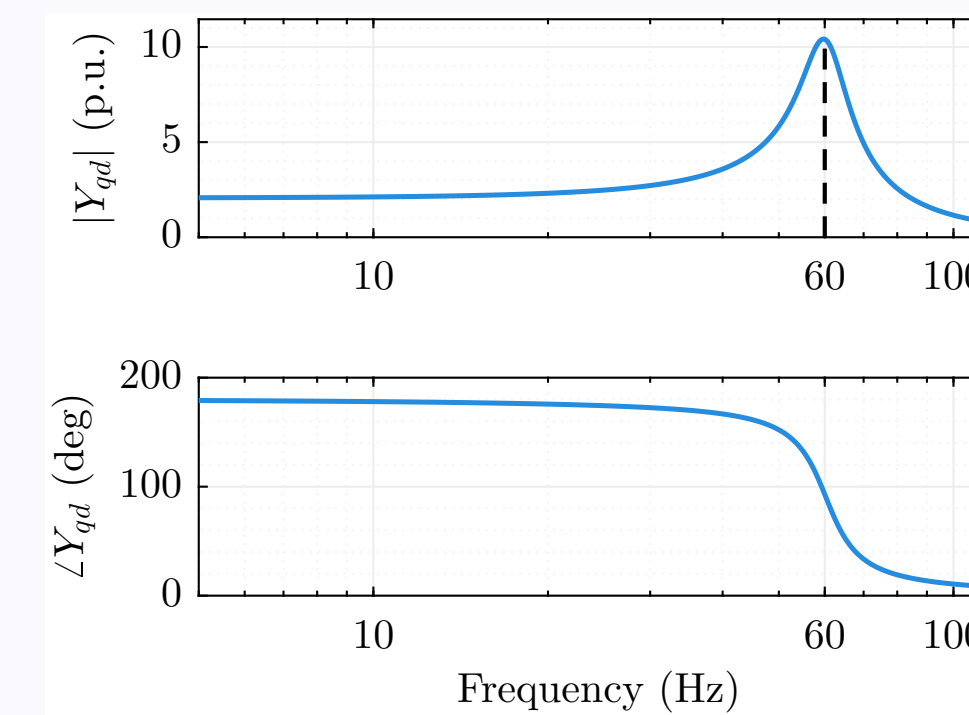


Figure 3: Y_{qd} for an ideal voltage source (HV defaults: $X_{Eff} = 0.48$ p.u., $X_{Eff}/R_{Eff} = 10$)

III. Equivalent Model

- Objective:**

$$\min_{R_{th}, L_{th}} \mathcal{J}(R_{th}, L_{th}) = \sqrt{\sum_{f_{min}}^{f_{max}} \frac{1}{2\pi f_k} \left(\frac{|Y_{qd}^{th}(f_k; R_{th}, L_{th})| - |Y_{qd}^{full}(f_k)|}{|Y_{qd}^{full}(f_k)|} \right)^2}$$

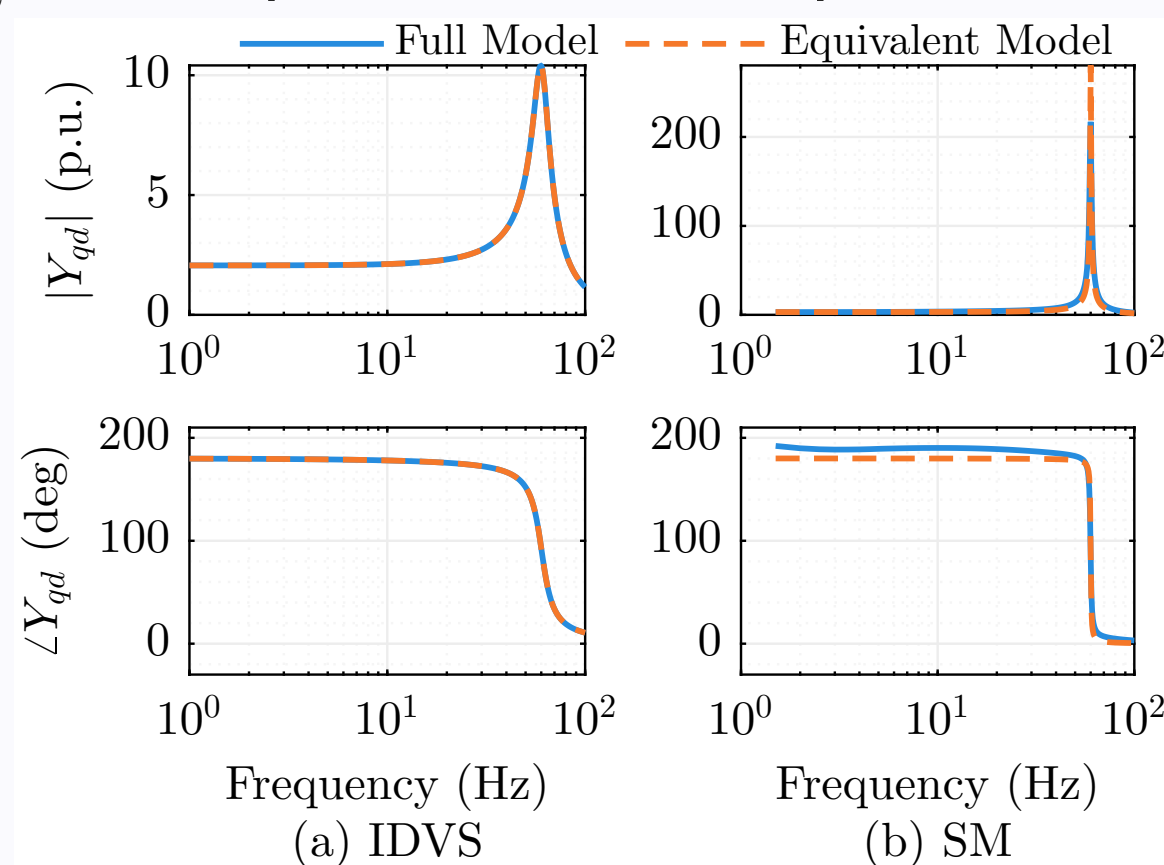
- $f_{min} = 5$ Hz $\rightarrow$ end of the transient time.
- $f_{max} = 100$ Hz $\rightarrow$ ignoring faster dynamics.
- More emphasis on lower frequencies.
- $|Y_{qd}|$ (p.u.) rather than Bode gain $\log_{10} |Y_{qd}|$ (dB).

$$(R_{Eff}, L_{Eff}) = \arg \min_{R_{th}, L_{th}} \mathcal{J}(R_{th}, L_{th})$$

- Compliance $\rightarrow (R_{Eff}, L_{Eff})$ within limits?

IV. Case Studies & Application

- Ideal Voltage Source $\rightarrow$ RMS error = 0.00%.
- SM $\rightarrow$ RMS error = 0.27%, accurately captures $x_d'' = 0.26$ p.u. & $r_a = 0.0025$ p.u..



- NREL GFM $\rightarrow$ RMS error = 0.037%, Fig. 5(b).**
- $(\omega_{res}^{GFM} = 58.01 \text{ Hz}) \leq (\omega_1 = 60 \text{ Hz})$.
- $X_{Eff}^{GFM} = 0.447 \therefore$ Compliant at HV terminals.

- Lower VCBW $\rightarrow \omega_{res} \downarrow, R_{Eff} \uparrow$, RMS error $\uparrow$, $\rightarrow$ weaker voltage-source behavior.**

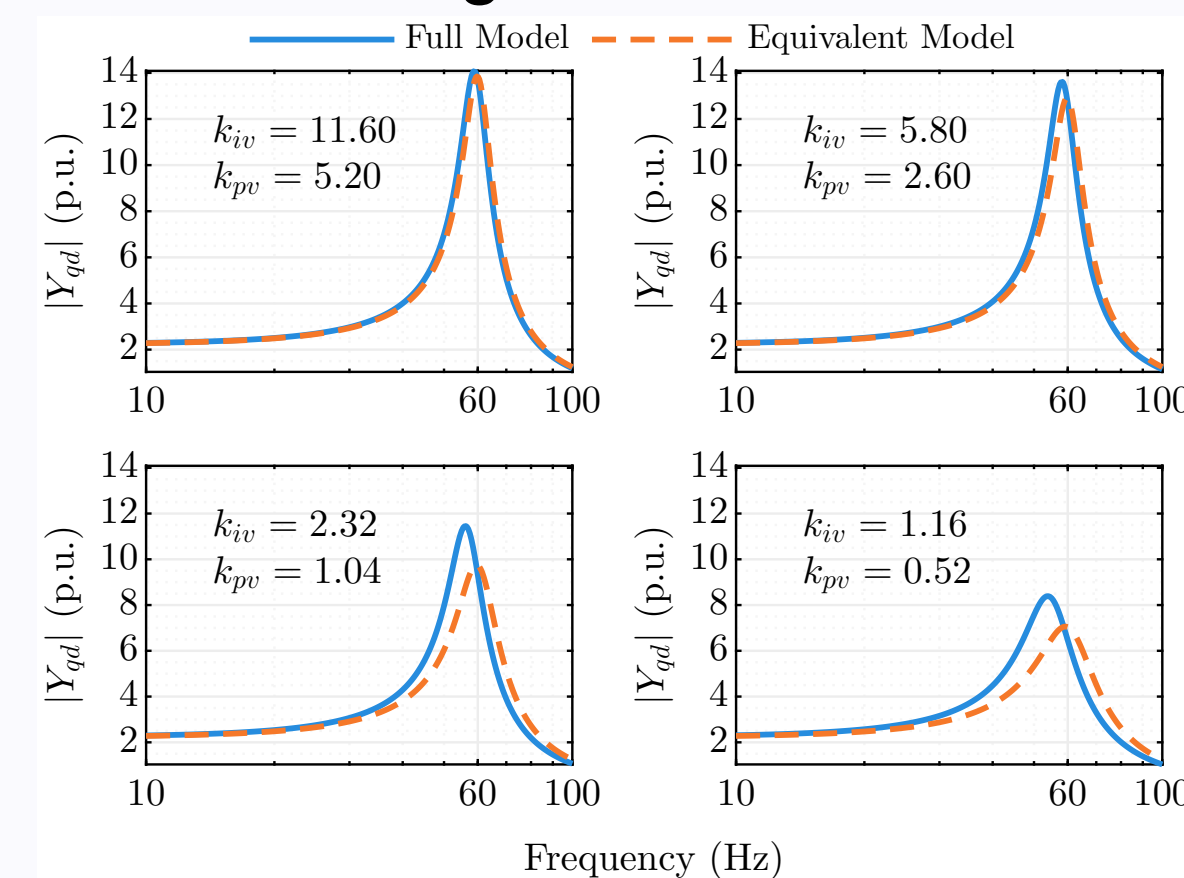


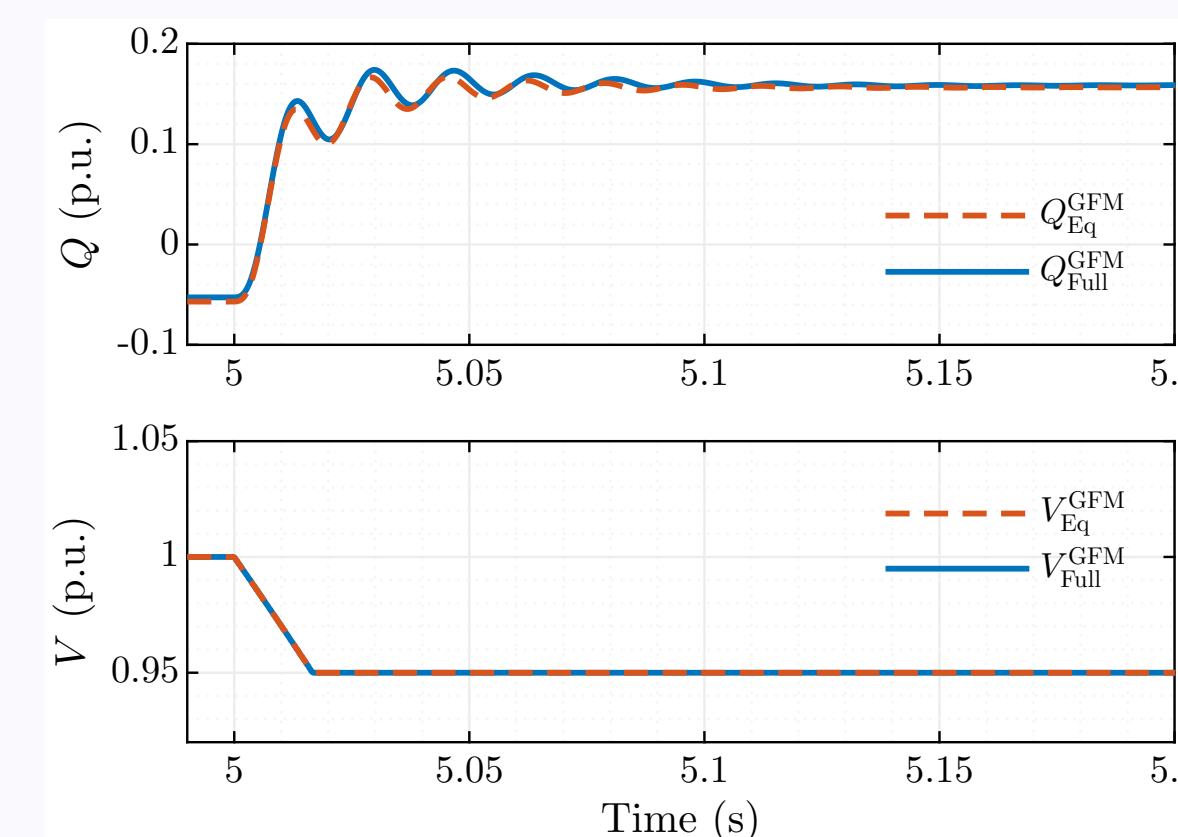
Figure 5: Y_{qd} admittance spectra for full vs equivalent models of GFM with changing voltage control gains

Table 2: Effect of changing voltage control loop gains on effective impedance (in p.u.) of the GFM

$\{k_{iv}, k_{pv}\}$ (p.u.)	ω_{res} (Hz)	X_{Eff}	R_{Eff}	RMS Error
{11.60, 5.20}	59.06	0.44	0.03	0.03%
{5.80, 2.60}	58.01	0.44	0.03	0.03%
{2.32, 1.04}	56.06	0.44	0.05	0.06%
{1.16, 0.52}	54.02	0.43	0.07	0.08%

- The equivalent model exhibits a similar reactive power response to voltage change.**

- Q_{Eq}^{GFM} vs Q_{Full}^{GFM} : RMS error (200 ms) = 2.76%.



- Likewise, fitting Y_{dq} would yield a similar active power response to voltage phase change.

- The equivalent model accurately captures the static voltage stability limit.**

- Avoids the need for EMT simulations.

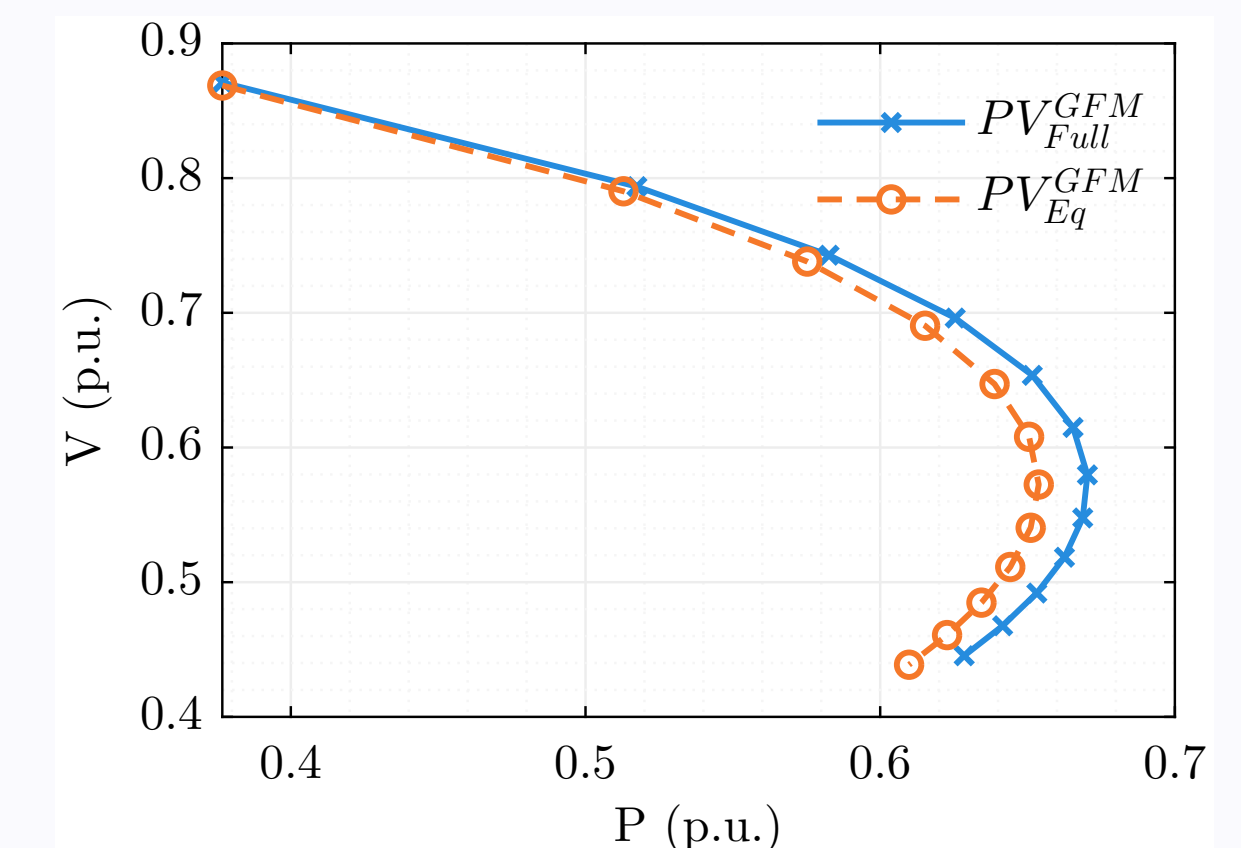


Figure 7: PV curve for full GFM model vs Thevenin equivalent

- $P_{Full}^{Max} = 0.67$ p.u. at $V_{Full}^{PMax} = 0.58$ p.u..
- $P_{Eq}^{Max} = 0.65$ p.u. at $V_{Eq}^{PMax} = 0.57$ p.u..
- Error in P^{Max} capability = 2.46%.